

# Single Input, Multiple Output DC-DC Buck Converter for Electric Vehicles

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**ABSTRACT** Increasing concerns on environmental pollution, global warming, depletion of the fossil fuel reserves and desire for reducing energy dependencies have led to an ever-increasing interest in electric vehicles (EV). The requirements for electric vehicles has brought many different problems and solutions in electric vehicle technology. One of these is the conversion of the voltage level from the battery in electric vehicles to other required voltage levels with DC-DC converters. As a solution, a separate converter can be used for each voltage level. Nevertheless, single-input multi-output (SIMO) converters can be used to reduce the cost and switching losses and hence improve system efficiency. In our study, we proposed non-isolated buck converter topology with single-input (48 V) and multi-output (12 V and 5 V). The 12 V voltage level is used for the horn, headlights while 5 V voltage level is used for the telemetry and microcontroller in electric vehicles. In this work, the general structure of a SIMO converter, design principles, small signal and stability analysis, and control steps are explained. The overall system has been simulated with Matlab / Simulink.

**Keywords:** Electric Vehicles, DC-DC Buck Converter, Single Input Multiple Output, Steady State Analyses, PI controller, Matlab/Simulink

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## 1. INTRODUCTION

Increasing demand for energy in the world and the diminishing of fossil energy sources promotes exploitation of other energy sources such as solar energy, fuel cells and other clean energy sources. These energies are usually environmentally friendly [1]. The primary utilizations of DC-DC converters include uninterruptible power supplies, battery charging/discharging devices, hybrid electric vehicle, and renewable energy systems [2-7]. Occasionally, in a typical buck converter, an active power switch replace the freewheeling diode [8-12]. A single input multiple-output DC-DC converter able of providing, boost, buck and inverted outputs at the same time has been presented in literature [13]. However, three switches are required for one output. This type of designs correspond only for low power application and output voltage. Multi-output DC-DC boost converter are studied in another study [14]. Output voltage for high and low

power applications is shared in this study. Nevertheless, for one output voltage two switches were required and its control scheme is complicated. Kumar and Omar present Single-Input Multiple-Output (SIMO) Synchronous DC-DC buck converter [15]. It has advantage of reducing the number of the switches; four switches are required for over three output voltage. Unfortunately, this SIMO converter has the disadvantage of requiring a higher current rating for four switches. Another single input, multiple outputs DC-DC converter has been proposed by Kwon, and Mora [16]. This converter is capable of providing both boost and inverted outputs. Nonetheless, in this new configuration, the loads are separately designed except the negative output. Double-output DC-DC buck converters with unidirectional and bidirectional characteristics has been presented by Santos [17]. However, the buck converter required power switches with high current ratings.

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This paper presents a bi-directional SIMO DC-DC buck converter for electric vehicles battery as shown in the Figure 1. The main aim of this converter is to buck the voltage coming from the battery to the different components. The presented topology has the advantage of having three switches for two outputs. In addition, the control strategies, design principles, and small signal analysis has been included. The system has been simulated in MATLAB/Simulink.

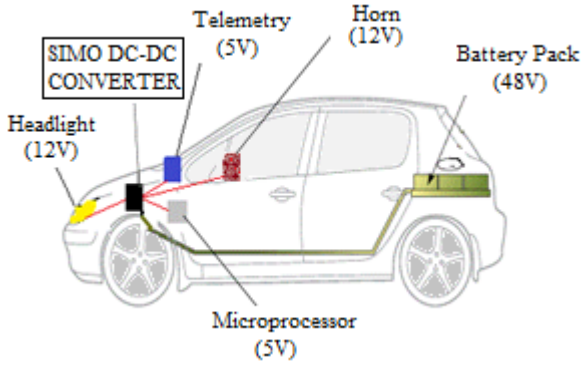


Fig. 1. SIMO DC-DC converter in electric vehicles

**2. BUCK AND SIMO CONVERTER TOPOLOGY**

**2.1. Buck Converter**

Buck converter is a step-down DC-to-DC converter, where the output voltage is lower than the input voltage [19]. The basic buck converter circuit is presented in Figure 2.

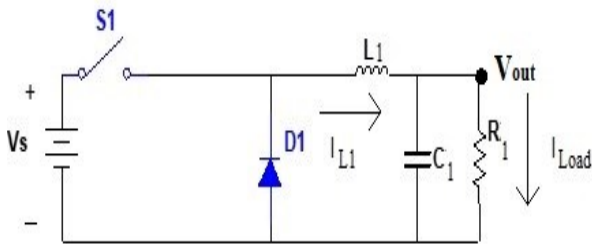


Fig. 2. Basic buck converter circuit

For the converter shown above, the current flows across the inductor in to the load when the switch ( $S_1$ ) is closed. This current charges the inductor ( $L_1$ ) by boosting both its magnetic field and voltage output. After a while, the output voltage ( $V_{out}$ ) will attain the desired value; then the switch ( $S_1$ ) is turned off and the current flows through the recovery diode ( $D_1$ ). At this state, inductor ( $L_1$ ) is discharged and current continues to flow through it. Before the inductor is fully discharged, the  $S_1$  is turned on,  $D_1$  is turned off and the cycle repeats. One can settle the ratio between the input and output voltage by modifying the duty cycle of the switch ( $S_1$ ).

**2.2. SIMO Converter**

Single input, multiple output topology that is used in this study is given in Figure 3.b. The bi-directional DC-DC converter used in this paper has less power loss distribution among the power switches than unidirectional characteristics [12]. The topology consists of three power switches  $S_1$ ;  $S_2$  and  $S_3$  as well as two low pass filters ( $L_1$ - $C_1$  and  $L_2$ - $C_2$ ) as illustrated in

Figure 3.b. The state of the switches is represented as *switch x = OFF (0)* and *switch x = ON*. Since there are three switches and two states for each switch, we obtain eight ways of operating of the presented converter [11]. Three switching states are operational, only. The other combinations were not included in this work. Table I presents the topological states (TS) of the system designed.

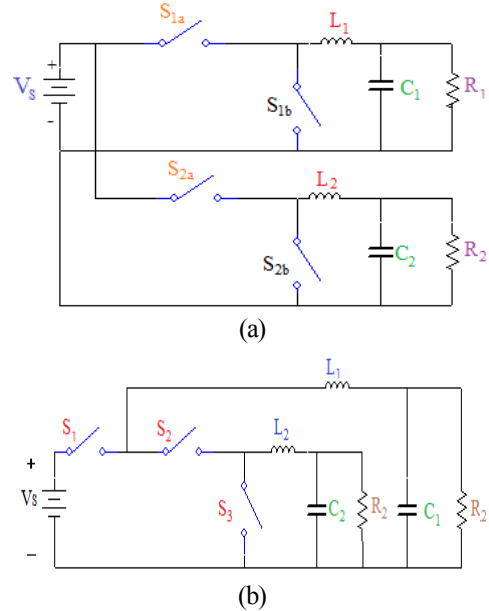


Fig. 3. Bi-directional SIMO DC-DC buck converter: (a) Parallel solution, (b) Implemented solution

The different switching states of SIMO converter are shown in Figure 4. It can be observed from the Figure 4 that:

- In state TS-1: Switch 1 = 1, Switch 2 = 1 and Switch 3 = 0. The input voltage ( $V_s$ ) supplies energy to the loads and to the inductors, in this state both  $L_1$  as well as  $L_2$  is charged.
- In state TS-2: Switch 1 = 1, Switch 2 = 0 and Switch 3 = 1. The input voltage ( $V_s$ ) supplies energy to  $R_1$ - $L_1$  and current of the inductor  $L_2$  ( $i_{L2}$ ) flows across  $S_2$ , delivering some of its energy to the load  $R_2$ . In this circumstance, inductance  $L_1$  and  $L_2$  will be respectively charged and discharged.
- In state TS-3: Switch 1 = 0, Switch 2 = 1 and Switch 3 = 1. The current in the inductor  $L_1$  ( $i_{L1}$ ) flows to both  $S_2$  and  $S_3$ , while  $i_{L2}$  flows only across  $S_3$ , delivering its stored energy to both loads  $R_1$  and  $R_2$ . In this situation the inductor  $L_1$ , as well as the inductor  $L_2$  are discharged.

Table 1. Topological states of the used SIMO converter

Topological states	TS-1	TS-2	TS-3
Switch 1	ON	ON	OFF
Switch 2	ON	OFF	ON
Switch 3	OFF	ON	ON

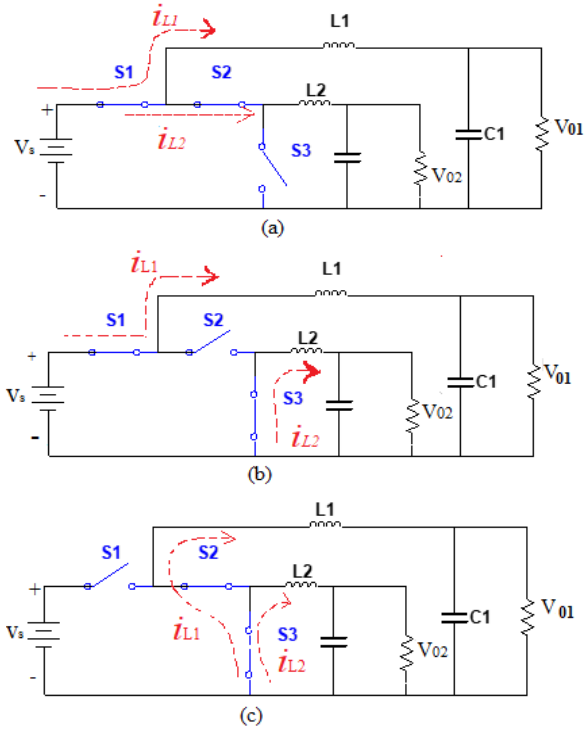


Fig. 4. The different switching states of the SIMO converter: (a) TS-1, (b) TS-2, (c) TS-3

### 3. STEADY STATE ANALYSIS

The presented SIMO converter is devised to operate in continuous conduction mode (CCM). The current and voltage waveforms of inductors are presented in Figure 5.  $T_{ON1}$  and  $T_{ON2}$  are the periods which the PWM generators one and two are generating the logic “1” at their corresponding outputs.

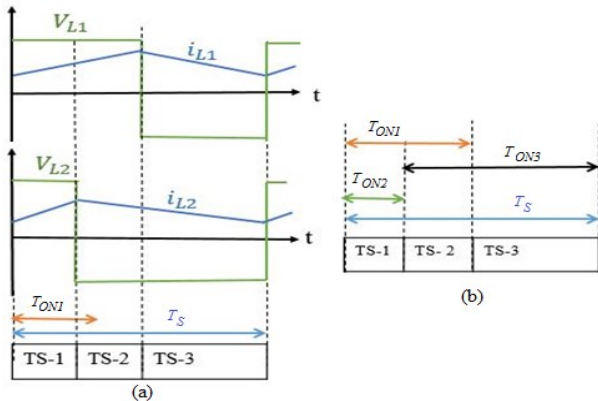


Fig. 5. Steady-state analysis of the SIMO converter; (a) Inductors current and voltage related with topological states, (b) Times in which the switch  $S_1$ ,  $S_2$  and  $S_3$  are turned ON.

From the topological states and waveforms found in the foregoing section, it can be inferred that the output voltage  $V_{01}$  and  $V_{02}$  controls voltage  $V_{R1}$  and  $V_{R2}$  respectively. Noting that the average inductor voltage is zero, the following equations are written:

$$(V_S - V_{01}) * T_{ON1} = V_{01} * (T_S - T_{ON1}) \quad (1)$$

$$(V_S - V_{02}) * (T_S - T_{ON3}) = V_{02} * T_{ON3} \quad (2)$$

According to the equations (1, 2), the different output voltages come out as a function of their input voltage and duty cycles.  $V_{01}$  and  $V_{02}$  are expressed in the following equations:

$$V_{01} = (D_1)V_S \quad (3)$$

$$V_{02} = (1 - D_2)V_S \quad (4)$$

where  $D_1$  and  $D_2$  represent the duty cycles of the system.

### 4. GENERALIZED STATE-SPACE AVERAGE MODEL

The voltage and the current dynamics are described by the state space average and it is given in (5)-(8):

$$L_1 \frac{di_{L1}}{dt} + V_{01} = D_1 V_S \quad (5)$$

$$C_1 \frac{dV_{01}}{dt} = i_{L1} - \frac{V_{01}}{R_1} \quad (6)$$

$$L_2 \frac{di_{L2}}{dt} + V_{02} = D_2 V_S \quad (7)$$

$$C_2 \frac{dV_{02}}{dt} = i_{L2} - \frac{V_{02}}{R_2} \quad (8)$$

By using these equations and the state equation, the transfer functions obtained between the output voltages and the duty ratios are as follows:

$$Tp_1(s) = \frac{V_S D_1 R_1}{(R_1 C_1 L_1 s^2 + L_1 s + R_1)} \quad (9)$$

$$Tp_2(s) = \frac{V_S D_2 R_2}{(R_2 C_2 L_2 s^2 + L_2 s + R_2)} \quad (10)$$

The circuit component's values are calculated as follow:

- Output 1: For  $P_1= 150W/12V$ ,  $V_S=48$ ,  $R_1=1\Omega$ ,  $L_1=180\mu F$ ,  $C_1=20\mu F$ .

- Output 2: For  $P_2= 25W/5V$ ,  $V_S=48$ ,  $R_2= 1\Omega$ ,  $L_2=89\mu F$ ,  $C_2=62\mu F$ .

In the open loop the circuit transfer function with their numerical values are found as follows:

$$Tp_1(s) = \frac{12}{(3.6*10^{-9}s^2 + 180*10^{-6}s + 1)} \quad (11)$$

$$Tp_2(s) = \frac{5}{(5.5*10^{-9}s^2 + 62*10^{-6}s + 1)} \quad (12)$$

### 5. SIMULATION AND RESULTS

#### 5.1. Simulink Model of SIMO DC-to-DC Converter

The simulation model of the whole system is shown in Figure 6. The system is composed of two PWM block, one logic block and the SIMO converter. The input voltage of the converter is 48 volt and the output is composed of 5 and 12 volts.

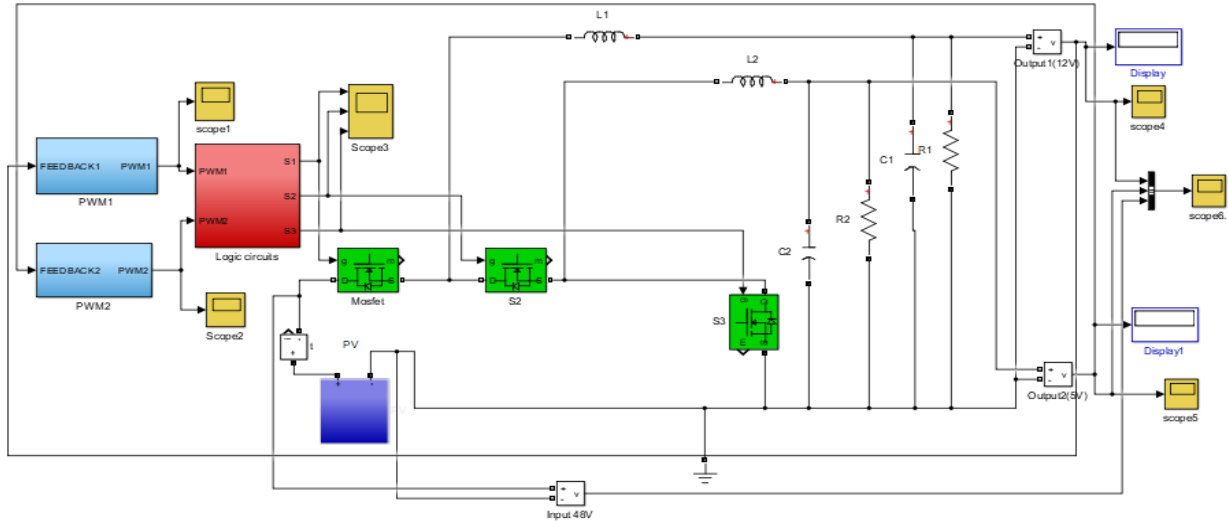


Fig. 6. Simulink diagram of the SIMO DC-DC converter

5.1.1. PWM Generator Implementation and PI Controller

All three switches discussed are controlled using PWM, and this block include PI controller. The Simulink model is given in Figure 7. DC-to-DC converters use switches to change the DC from one level to another [20]. The system operates at 50 kHz with an output value between 0 and 1. The PWM1 and PWM2 generators produce an error signal and inserted into the PI block. The output of the PI block is compared to the sawtooth. Thus, logic 1 and 0 values are produced. The PI controller is a proportional-integral controller. In our case, we used PI to control the output voltage coming from the SIMO DC-DC buck converter [18, 21]. The values of the gains  $K_p$ , and  $K_i$ , are chosen carefully using Routh–Hurwitz stability criterion. The output would reach the reference value with a very short settling time and without an overshoot.

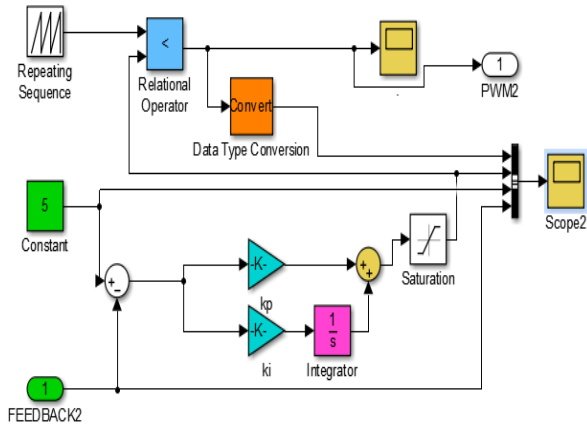


Fig. 7. PWM and PI design with Matlab/Simulink

5.1.2 Modelling the Logic Circuits

The outputs of the PWM1 and PWM2 which is illustrated in Figure 7 straightly go through a logic circuit (LC) block as given in Figure 6. This LC block command the state of the switches  $S_1$ ,  $S_2$  and  $S_3$ . From the topological states point of view (shown in Table I), it can be deduced that the charging of the inductor  $L_1$  is

fully reliant to switch 1 ( $S_1$ ), consequently to control  $V_{O1}$  the PWM1 is directly connected to  $S_1$ . In contrary, either charging or discharging of  $L_2$  is not reliant on the state of  $S_2$  since when the  $L_2$  is discharging in the TS-3 the switch  $S_2$  is ON. The second PWM generator defines an interval that the switch  $S_2$  should begin to operate. In addition, switch  $S_2$  should be controlled in such a way that prohibited states are avoided. These requirements are achieved by using a two input “OR” gate and a “NOT” gate shown. Lastly, the other two switches specify the switch  $S_3$ . This deduces that the only work for the switch  $S_3$  to avoid the prohibited states. This is achieved using NAND gate. The logic block is shown in Figure 8(a) and the gate signals of  $S_1$ ,  $S_2$  and  $S_3$  are presented in Figure 8(b).

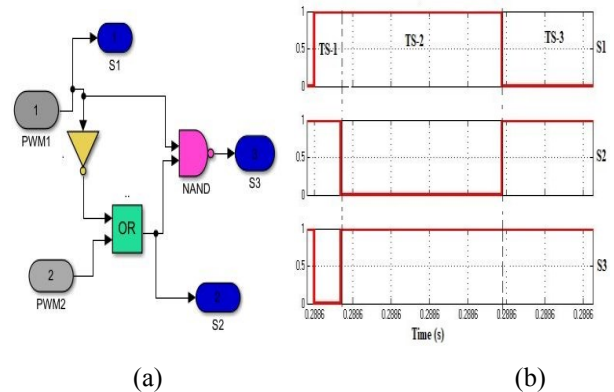


Fig. 8. Gate signals of the switches: (a) Logic block; (b) Output signals of the block

5.2. SIMO DC-DC Converter Represented with Transfer Function

The simulation diagram of single input, multiple output (SIMO) converter along with transfer function presented in this paper and the results are shown respectively in Figures (9, 10).



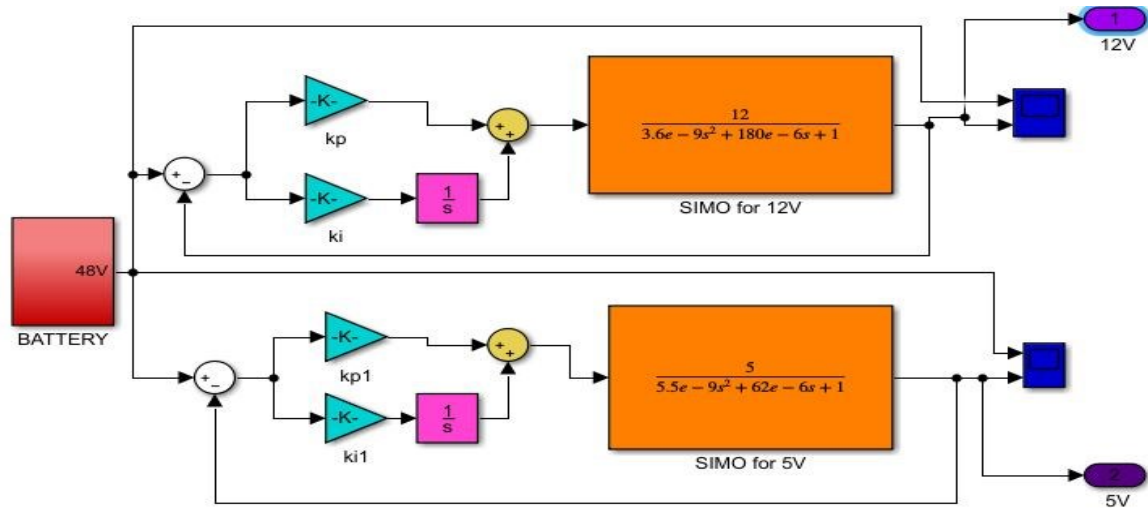


Fig. 9. SIMO converter with transfer function diagram

Figure 10 shows the simulation results of the transfer function of DC-DC converter. It can be deduced that all outputs reach their values in short time.

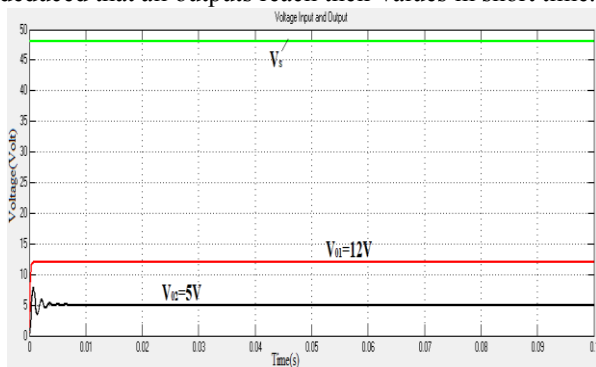


Fig. 10. The SIMO DC-DC converter

The PI controller continuously detect an error signal as the difference between feedback loop and the reference voltage. During the first zone, the feedback signal (black line) is less than the reference value (green line) as seen in the Figure 11, so error is negative. The PI controller produces signal (red color line) to eliminate the difference between feedback and reference and the switch  $S_1$  is ON. During this period, the PWM generate pulses. In the second section, the feedback is higher than reference of 12 V, so error is positive. The PI is not generating a signal and the switch  $S_1$  is OFF.

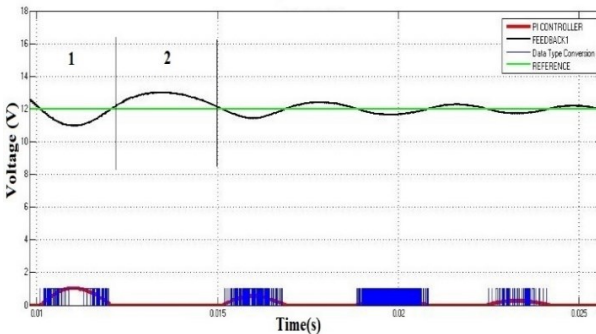
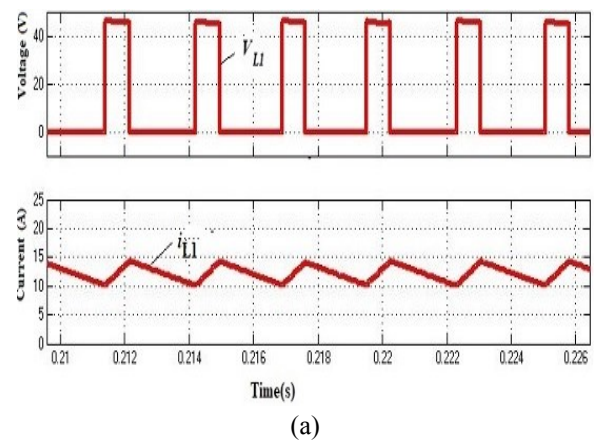
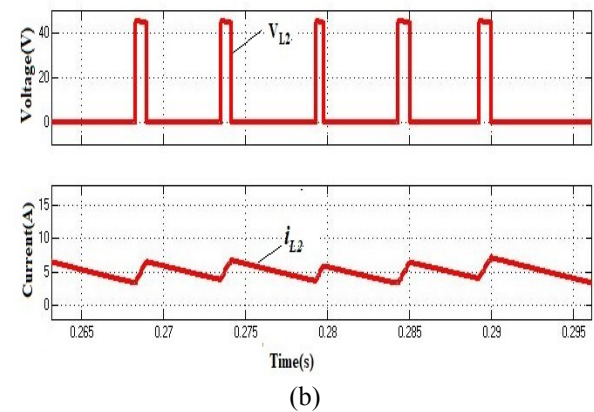


Fig. 11. Working principle of PI controller designed with Matlab/Simulink



(a)



(b)

Fig. 12. Voltage and current associated with each inductor: (a) Variables of the inductor  $L_1$ , (b) Variables of the inductor  $L_2$

The Simulink model of the SIMO DC-DC converter is given in Figure 13. Adding that all the required output voltages reaches their values with a very small overshoot.

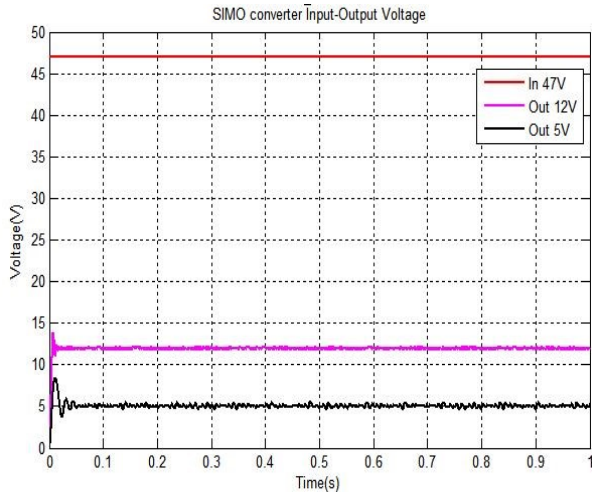


Fig. 13. Input and output voltages of SIMO DC-DC converter

Current in the load  $R_1$  and  $R_2$  is shown respectively in Figure 14 (a, b) when both loads are equal to  $1\Omega$ . When load is increased to  $R_1 = R_2 = 1.5\Omega$ , one can observe that the output current decrease as shown in Figure 15 (a, b). This implies that the output currents depend on the load. When the load is increased, the current decreases. The same when the load decreases, the current increases.

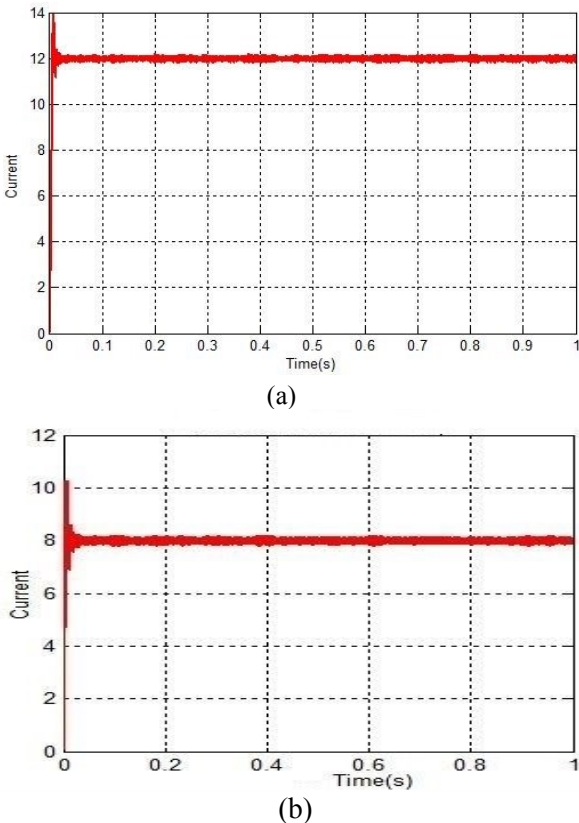


Fig. 14. The current as a function of loads: (a) Load  $R_1=1\Omega$  (b), Load  $R_2=1\Omega$

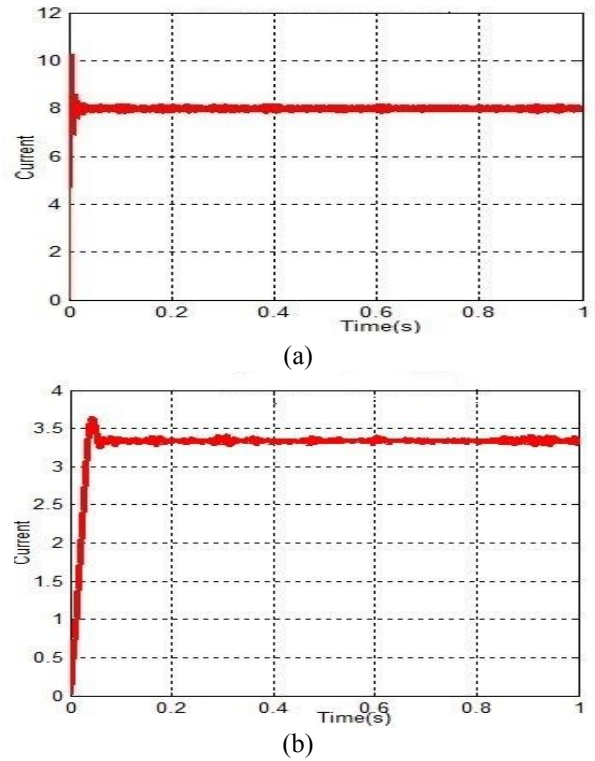


Fig. 15. The current as a function of loads: (a) Load  $R_1=1.5\Omega$  (b), Load  $R_2=1.5\Omega$

## 6. CONCLUSION

In this study, we have used single-input, multiple-output converter that can be used for electric vehicles. Steady state analysis of the models, control strategies and PWM implementations are presented. The simulation of the both dynamic model and static model is performed in Matlab/Simulink. Simulink results show that this converter is suitable for the electrical vehicles. The idea of using only one converter that can supply power for different voltage output requirements is not far away. Efforts continue to realize single-input multiple-output buck converter more efficient in their control schemes and cost-effective for the integration of electric vehicles, renewable energy systems, mobile equipments, batteries and other applications.

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